

APPLICATIONS OF LINEAR SYSTEMS

▼ DC circuits

- Every circuit component is associated with a voltage drop across that component.
- The circuit components we are interested are:

Name	Notation	$V_{AB} = V_A - V_B$
Generator		$V_{AB} = -E$
Resistor		$V_{AB} = IR$
Inductor		$V_{AB} = L \frac{dI}{dt}$
Capacitor		$V_{AB} = \frac{1}{C} \int_{-\infty}^t I d\tau$

V_{AB} = voltage drop from A to B

E = voltage of DC generator

R = resistance

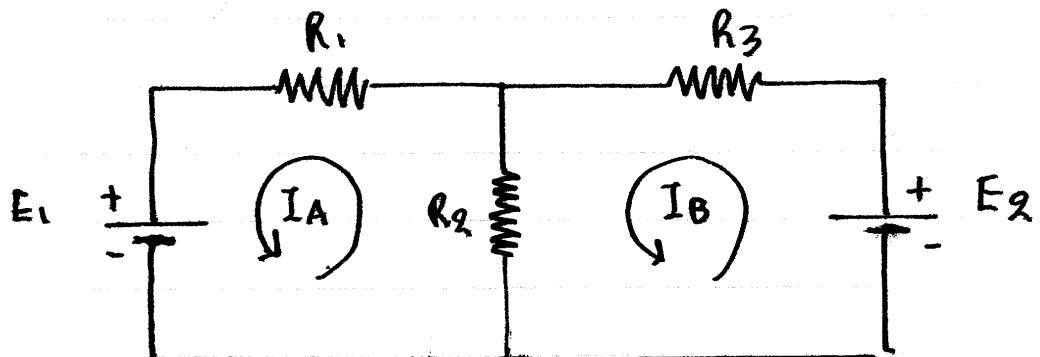
L = inductance

C = capacitance.

► Method: To calculate the currents around a circuit we work as follows.

- ₁ Define loop currents $I_A, I_B, \text{etc.}$ associated with each loop.
- ₂ For each, the sum all of all voltage drops around the loop must be zero.
Thus, for each loop, we have an equation.
- ₃ Solve the system of equations to find the loop currents.
- ₄ From the loop currents we may calculate the branch currents, voltage drops, etc.

example



$$\text{Loop A: } \{ +E_1 + I_A R_1 + (I_A - I_B) R_2 = 0 \leftrightarrow$$

$$\text{Loop B: } \{ -E_2 + I_B R_3 + (I_B - I_A) R_2 = 0$$

$$\leftrightarrow \begin{cases} (R_1 + R_2) I_A + (-R_2) I_B = -E_1 \\ (-R_2) I_A + (R_2 + R_3) I_B = E_2 \end{cases}$$

$$\Leftrightarrow \underbrace{\begin{bmatrix} R_1 + R_2 & -R_2 \\ -R_2 & R_2 + R_3 \end{bmatrix}}_A \begin{bmatrix} I_A \\ I_B \end{bmatrix} = \begin{bmatrix} -E_1 \\ E_2 \end{bmatrix}$$

$$\begin{aligned} \det A &= \begin{vmatrix} R_1 + R_2 & -R_2 \\ -R_2 & R_2 + R_3 \end{vmatrix} = \\ &= (R_1 + R_2)(R_2 + R_3) - (-R_2)^2 = \\ &= R_1 R_2 + R_1 R_3 + R_2^2 + R_2 R_3 - R_2^2 = \\ &= R_1 R_2 + R_2 R_3 + R_3 R_1 \neq 0 \\ &\text{since } R_1 > 0, R_2 > 0 \text{ and } R_3 > 0. \end{aligned}$$

Thus:

$$\begin{bmatrix} I_A \\ I_B \end{bmatrix} = \frac{1}{\det A} \begin{bmatrix} R_2 + R_3 & +R_2 \\ +R_2 & R_1 + R_2 \end{bmatrix} \begin{bmatrix} -E_1 \\ E_2 \end{bmatrix}$$

$$= \frac{1}{\det A} \begin{bmatrix} -E_1(R_2 + R_3) + E_2 R_2 \\ -E_1 R_2 + E_2 (R_1 + R_2) \end{bmatrix}$$

so

$$I_A = \frac{-E_1(R_2 + R_3) + E_2 R_2}{R_1 R_2 + R_2 R_3 + R_3 R_1}$$

$$I_B = \frac{-E_1 R_2 + E_2 (R_1 + R_2)}{R_1 R_2 + R_2 R_3 + R_3 R_1}$$

- Current through $R_1 : I_A$
 $R_2 : I_A - I_B$
 $R_3 : I_B$.

- Find the necessary and sufficient condition so there is no current through R_2 .

Balance $\Leftrightarrow I_A - I_B = 0 \Leftrightarrow I_A = I_B \Leftrightarrow$

$$\Leftrightarrow -E_1(R_2 + R_3) + E_2 R_2 = E_1 R_2 + E_2(R_1 + R_2)$$

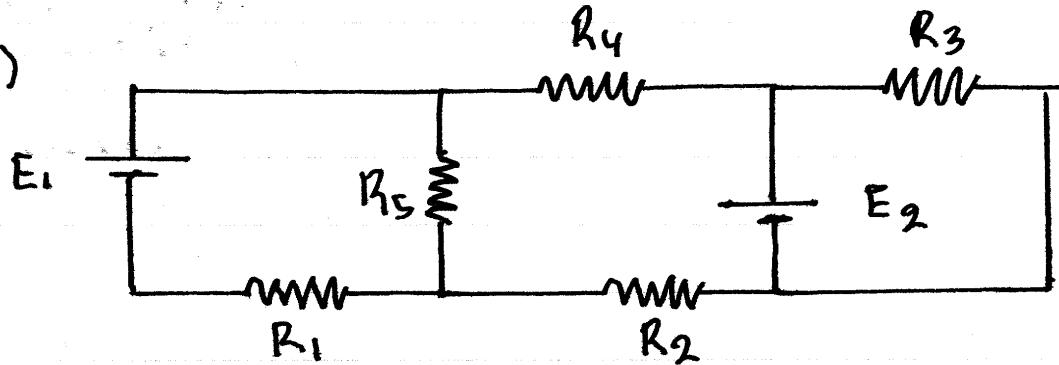
$$\Leftrightarrow \underline{E_1 R_2} + \underline{E_1 R_3} - \underline{E_2 R_2} = \underline{E_1 R_2} - \underline{E_2 R_1} - \underline{E_2 R_2}$$

$$\Leftrightarrow E_1 R_3 = -E_2 R_1 \Leftrightarrow \boxed{E_1 = -E_2 \frac{R_1}{R_3}}$$

EXERCISES

- ① Find the loop currents in the following circuits.

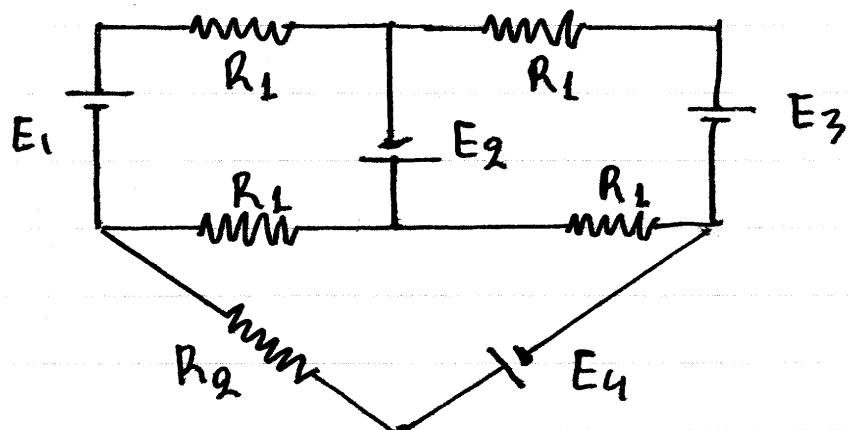
a)

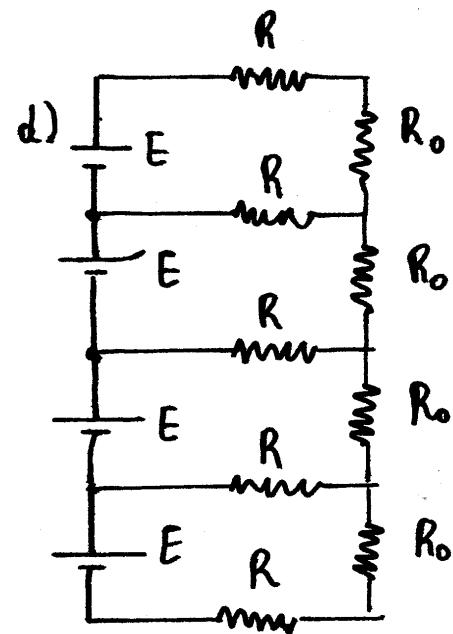
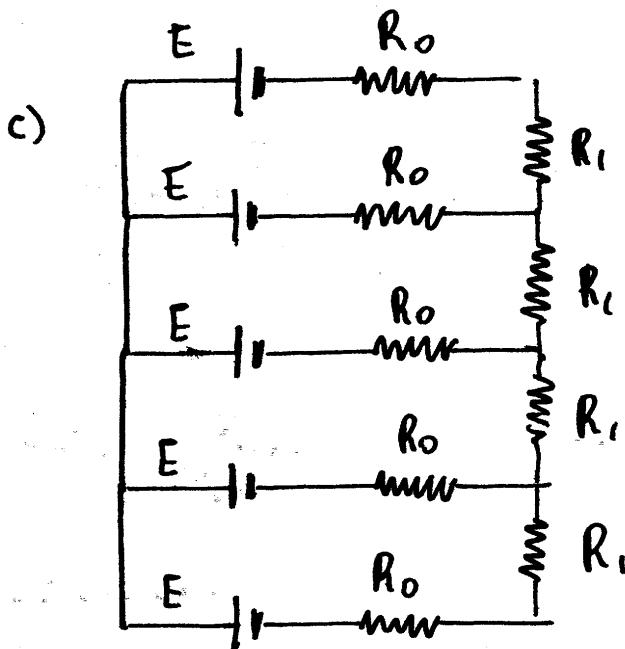


$$\text{when } R_1 = R_2 = R_3 = a$$

$$R_4 = R_5 = b$$

b)

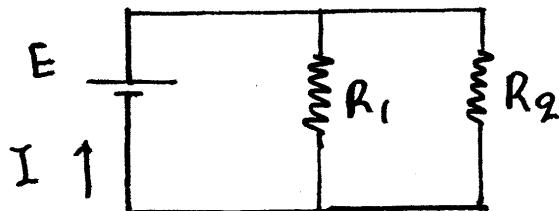




(2) Parallel Resistors.

Show that two resistors R_1, R_2 connected in parallel to a generator are equivalent to a single resistor R with

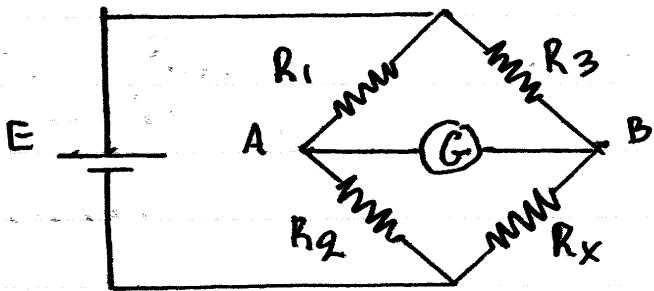
$$\frac{1}{R} = \frac{1}{R_1} + \frac{1}{R_2} \quad (1)$$



(i.e. $E = IR$
with R given
by (1)).

③ Wheatstone Bridge

The Wheatstone bridge can be used to measure the unknown resistance R_x .



⑥ is a galvanometer measuring voltage drop V_{AB} between A and B.

Assume the Galvanometer has resistance R_g .

- Solve for the 3 loop currents in general.
- Show that when $V_{AB} = 0$ then

$$R_x = \left(\frac{R_2}{R_1} \right) R_3.$$

→ Superposition principle

- Any general DC circuit with n loops and m generators gives a linear system of equations of the form

$$R \mathbf{J} = P \mathbf{E}$$

with R the resistance matrix ($R \in M_{nn}(R)$)
 P the source configuration matrix
 $(P \in M_{nm}(R))$

and

$$\mathbf{J} = \begin{bmatrix} I_1 \\ I_2 \\ \vdots \\ I_n \end{bmatrix} \text{ and } \mathbf{E} = \begin{bmatrix} E_1 \\ E_2 \\ \vdots \\ E_m \end{bmatrix}$$

- If R has an inverse then this system has a unique solution

$$\mathbf{J} = R^{-1} P \mathbf{E}$$

- Thm : Let \mathbf{J}_1 be the solution when all generators are turned off except E_1 , and similarly for $\mathbf{J}_2, \mathbf{J}_3, \dots, \mathbf{J}_n$
 Then

$$\mathbf{J} = \mathbf{J}_1 + \mathbf{J}_2 + \dots + \mathbf{J}_n$$

Proof

Define

$$\boldsymbol{\varepsilon}_1 = \begin{bmatrix} E_1 \\ 0 \\ \vdots \\ 0 \end{bmatrix}, \quad \boldsymbol{\varepsilon}_2 = \begin{bmatrix} 0 \\ E_2 \\ \vdots \\ 0 \end{bmatrix}, \dots, \boldsymbol{\varepsilon}_m = \begin{bmatrix} 0 \\ 0 \\ \vdots \\ E_m \end{bmatrix}$$

Then

$$\boldsymbol{\varepsilon} = \boldsymbol{\varepsilon}_1 + \boldsymbol{\varepsilon}_2 + \dots + \boldsymbol{\varepsilon}_m$$

For each problem we get a linear system.

Altogether:

$$\left\{ \begin{array}{l} Q \mathbf{f}_1 = P \boldsymbol{\varepsilon}_1 \\ Q \mathbf{f}_2 = P \boldsymbol{\varepsilon}_2 \\ \vdots \\ Q \mathbf{f}_m = P \boldsymbol{\varepsilon}_m \end{array} \right. \Rightarrow \left\{ \begin{array}{l} \mathbf{f}_1 = Q^{-1} P \boldsymbol{\varepsilon}_1 \\ \mathbf{f}_2 = Q^{-1} P \boldsymbol{\varepsilon}_2 \\ \vdots \\ \mathbf{f}_m = Q^{-1} P \boldsymbol{\varepsilon}_m \end{array} \right.$$

It follows that

$$\begin{aligned} \mathbf{f} &= Q^{-1} P \boldsymbol{\varepsilon} = Q^{-1} P (\boldsymbol{\varepsilon}_1 + \boldsymbol{\varepsilon}_2 + \dots + \boldsymbol{\varepsilon}_m) \\ &= Q^{-1} P \boldsymbol{\varepsilon}_1 + Q^{-1} P \boldsymbol{\varepsilon}_2 + \dots + Q^{-1} P \boldsymbol{\varepsilon}_m \\ &= \mathbf{f}_1 + \mathbf{f}_2 + \dots + \mathbf{f}_m. \end{aligned}$$

→ Incidence Matrices

- Consider a circuit with n -loops. Define
 E_{ab} = the total generators shared by
 loop a and b

R_{ab} = the total resistance shared by
 loop a and b

I_a = the loop current for loop a

$$\delta_{ab} = \begin{cases} 1, & \text{if } a=b \\ 0, & \text{if } a \neq b. \end{cases}$$

and note that $E_{ab} = E_{ba}$ and $R_{ab} = R_{ba}$

- Can we write R_{ab} in terms of R_{ab} ?

- Note that the equation for loop a reads:

$$\sum_b E_{ab} = \sum_{b \neq a} R_{ab}(I_a - I_b) + R_{aa}I_a$$

The right-hand-side can be rewritten as:

$$RHS = \sum_{b \neq a} R_{ab}(I_a - I_b) + R_{aa}I_a$$

$$= I_a \sum_b R_{ab} - \sum_{b \neq a} R_{ab}I_b$$

$$= I_a \sum_y R_{ay} - \sum_b R_{ab}(1 - \delta_{ab})I_b =$$

$$= \sum_b \left[\delta_{ab} \sum_y R_{by} \right] I_b - \sum_b R_{ab}(1 - \delta_{ab})I_b =$$

$$= \sum_b \left[\delta_{ab} \sum_y R_{by} - R_{ab} (1 - \delta_{ab}) \right] I_b =$$

$$= \sum_b R_{ab} I_b$$

therefore

$$R_{ab} = \delta_{ab} \sum_y R_{by} - R_{ab} (1 - \delta_{ab})$$

example

For $n=2$: two loops

$$R_{11} = \delta_{11} \sum_y R_{1y} - R_{11} (1 - \cancel{\delta_{11}}) = R_{11} + R_{12}$$

$$R_{12} = \delta_{12} \sum_y R_{1y} - R_{12} (1 - \delta_{12}) = -R_{12}$$

$$R_{21} = \cancel{\delta_{21}} \sum_y R_{2y} - R_{21} (1 - \cancel{\delta_{21}}) = -R_{21} = -R_{12}$$

$$R_{22} = \delta_{22} \sum_y R_{2y} - R_{22} (1 - \cancel{\delta_{22}}) = R_{21} + R_{22}$$

$$= R_{12} + R_{22}$$

thus $R = \begin{bmatrix} R_{11} + R_{12} & -R_{12} \\ -R_{12} & R_{22} + R_{12} \end{bmatrix}$

- It is not always true that R has an inverse.
e.g. when $R_{11}=0$ and $R_{22}=0$
then $\det(R)=0$
This situation indicates a short-circuit.
- The incidence matrix can be of use if you want to write a computer program that can solve arbitrary circuits.

EXERCISE

- ④ Derive the relation between R and the incidence matrix R for the general $n=3$ circuit (3 loops). Then calculate and simplify the determinant $\det(R)$. Under what conditions is R singular?

I Least-squares fit

- Consider a set of data points

$$(x_1, y_1), (x_2, y_2), \dots, (x_n, y_n)$$

that approximately fall upon a line

$$(l): y = ax + b.$$

- Want to find the best possible values for a and b .

► Solution

Define $\sum x = x_1 + x_2 + \dots + x_n$

$$\sum y = y_1 + y_2 + \dots + y_n$$

$$\sum x^2 = x_1^2 + x_2^2 + \dots + x_n^2$$

$$\sum y^2 = y_1^2 + y_2^2 + \dots + y_n^2$$

and $\sum xy = x_1y_1 + x_2y_2 + \dots + x_ny_n$.

We estimate the error in the line fit by calculating

$$E(a, b) = \sum_{k=1}^n (y_k - (ax_k + b))^2$$

To minimize $E(a, b)$ we calculate the partial derivatives with respect to a and b :

$$\begin{aligned}
 \frac{\partial E(a, b)}{\partial a} &= \frac{\partial}{\partial a} \sum_{k=1}^n [y_k - (ax_k + b)]^2 = \\
 &= \sum_{k=1}^n \left\{ \frac{\partial}{\partial a} [y_k - (ax_k + b)]^2 \right\} = \\
 &= \sum_{k=1}^n \left\{ 2[y_k - (ax_k + b)](-x_k) \right\} = \\
 &= \sum_{k=1}^n [-2x_k y_k + 2ax_k^2 + 2bx_k] \\
 &= -2 \sum_{k=1}^n x_k y_k + 2a \sum_{k=1}^n x_k^2 + 2b \sum_{k=1}^n x_k \\
 &= -2 \sum x_y + 2a \sum x_x + 2b \sum x
 \end{aligned}$$

and

$$\begin{aligned}
 \frac{\partial E(a, b)}{\partial b} &= \frac{\partial}{\partial b} \sum_{k=1}^n [y_k - (ax_k + b)]^2 = \\
 &= \sum_{k=1}^n \left\{ \frac{\partial}{\partial b} [y_k - (ax_k + b)]^2 \right\} = \\
 &= \sum_{k=1}^n \left\{ 2[y_k - (ax_k + b)](-1) \right\} = \\
 &= \sum_{k=1}^n [-2y_k + 2ax_k + 2b] \\
 &= -2 \sum_{k=1}^n y_k + 2a \sum_{k=1}^n x_k + 2nb \\
 &= -2 \sum y + 2a \sum x + 2nb
 \end{aligned}$$

At the minimum we have

$$\left\{ \begin{array}{l} \frac{\partial E(a, b)}{\partial a} = 0 \\ \frac{\partial E(a, b)}{\partial b} = 0 \end{array} \right. \Leftrightarrow \left\{ \begin{array}{l} -2S_{xy} + 2aS_{xx} + 2bS_x = 0 \\ -2S_y + 2aS_x + 2b = 0 \end{array} \right.$$

$$\Leftrightarrow \begin{bmatrix} S_{xx} & S_x \\ S_x & n \end{bmatrix} \begin{bmatrix} a \\ b \end{bmatrix} = \begin{bmatrix} S_{xy} \\ S_y \end{bmatrix}$$

$$D_a = \begin{vmatrix} S_{xx} & S_x \\ S_x & n \end{vmatrix} = nS_{xx} - [S_x]^2$$

$$D_b = \begin{vmatrix} S_{xy} & S_x \\ S_y & n \end{vmatrix} = nS_{xy} - S_x S_y$$

$$D_b = \begin{vmatrix} S_{xx} & S_{xy} \\ S_x & S_y \end{vmatrix} = S_{xx} S_y - S_{xy} S_x$$

consequently,

$$a = \frac{nS_{xy} - S_x S_y}{nS_{xx} - [S_x]^2}$$

$$b = \frac{S_{xx} S_y - S_{xy} S_x}{nS_{xx} - [S_x]^2}$$

• Will we always have

$$D = n \sum_{k=1}^n x_k^2 - [\sum_{k=1}^n x_k]^2 \neq 0 ?$$

Answer: We use the Lagrange identity:

$$\left(\sum_{k=1}^n a_k^2 \right) \left(\sum_{k=1}^n b_k^2 \right) - \left(\sum_{k=1}^n a_k b_k \right)^2 = \\ = \frac{1}{2} \sum_{k=1}^n \sum_{l=1}^n \begin{vmatrix} a_k & b_k \\ a_l & b_l \end{vmatrix}^2$$

It follows that

$$D = n \sum_{k=1}^n x_k^2 - [\sum_{k=1}^n x_k]^2 = \\ = (1^2 + 1^2 + \dots + 1^2)(x_1^2 + x_2^2 + \dots + x_n^2) - \\ - (\sum_{k=1}^n x_k)^2 = \\ = \frac{1}{2} \sum_{k=1}^n \sum_{l=1}^n \begin{vmatrix} 1 & x_k \\ 1 & x_l \end{vmatrix}^2 = \\ = \frac{1}{2} \sum_{k=1}^n \sum_{l=1}^n (x_k - x_l)^2 \geq 0$$

As long as at least two of x_1, x_2, \dots, x_n are not equal to each other, we see that $D > 0 \Rightarrow D \neq 0$.

- How do we know that the solution we found is really a minimum and not a maximum?

Answer: A sufficient condition for a minimum is that

$$\frac{\partial^2 E(a, b)}{\partial a^2} > 0 \quad (1) \text{, and}$$

$$M(a, b) = \frac{\partial^2 E(a, b)}{\partial a^2} \frac{\partial^2 E(a, b)}{\partial b^2} - \left[\frac{\partial^2 E(a, b)}{\partial a \partial b} \right]^2 > 0 \quad (2)$$

(1) \rightarrow minimum in a direction

(2) \rightarrow some curvature in all directions.

Recall that

$$\frac{\partial E(a, b)}{\partial a} = -2 \varsigma_{xy} + 2a \varsigma'_{xx} + 2b \varsigma_x$$

$$\frac{\partial E(a, b)}{\partial b} = -2 \varsigma_y + 2a \varsigma_x + 2nb$$

It follows that

$$\frac{\partial^2 E(a, b)}{\partial a^2} = 2 \varsigma_{xx}, \quad \frac{\partial^2 E(a, b)}{\partial b^2} = 2n, \quad \text{and}$$

$$\frac{\partial^2 E(a, b)}{\partial a \partial b} = 2 \varsigma'_x$$

and thus for eq. (1):

$$\frac{\partial^2 E(a, b)}{\partial a^2} = 2 \sum_{k=1}^n x_k^2 > 0$$

if at least one $x_k \neq 0$.

$$M(a, b) = (2 \sum_{k=1}^n x_k)(2n) - (2 \sum_{k=1}^n x_k)^2 =$$

$$!!! \left(= 4(n \sum_{k=1}^n x_k - (\sum_{k=1}^n x_k)^2) \right)$$

$$= 4D > 0 \leftarrow \text{if at least one } x_k - x_l \neq 0 \text{ for } k \neq l.$$

► Note the amusing relationship between $M(a, b)$ and the determinant D !!

EXERCISES

⑤ Find the least-square fit line given the data

a) $(-1, -2), (0, 0), (1, 2 + \Delta)$

b) $(0, 0), (1, 1), (2, 2 + \varepsilon)$

c) $(1, 1), (a, a), (2, 2 + a)$

⑥ Confirm that given two data points (x_1, y_1) and (x_2, y_2) the least-square fit line corresponds to the line that does in fact go through the two data points (i.e. you have a perfect fit). You will find

$$a = \frac{y_1 - y_2}{x_1 - x_2} \leftarrow \text{slope}$$

$$b = y_1 - ax_1$$

which corresponds to the line

$$(l): y - y_1 = a(x - x_1)$$

⑦ Suppose that your data points are not exact and you are given instead

$$(x_k \pm \sigma_{x_k}, y_k \pm \sigma_{y_k}), k=1, 2, 3, \dots, n$$

You may still calculate a, b as usual but because your data is not exact there will be some error in your predicted a and b . So we want $a \pm \sigma_a$, and $b \pm \sigma_b$.

It can be shown that

$$\sigma_a^2 = \sum_{k=1}^n \left(\frac{\partial a}{\partial x_k} \right)^2 \sigma_{x_k}^2 + \sum_{k=1}^n \left(\frac{\partial a}{\partial y_k} \right)^2 \sigma_{y_k}^2$$

$$\sigma_b^2 = \sum_{k=1}^n \left(\frac{\partial b}{\partial x_k} \right)^2 \sigma_{x_k}^2 + \sum_{k=1}^n \left(\frac{\partial b}{\partial y_k} \right)^2 \sigma_{y_k}^2$$

Calculate the derivatives

$$\frac{\partial a}{\partial x_k}, \frac{\partial a}{\partial y_k}, \frac{\partial b}{\partial x_k}, \frac{\partial b}{\partial y_k}$$

→ Obviously, given these derivatives we can write a computer program that finds σ_a and σ_b in addition to a and b from the data.